Compressive Sensing Part V: Beyond Recovery

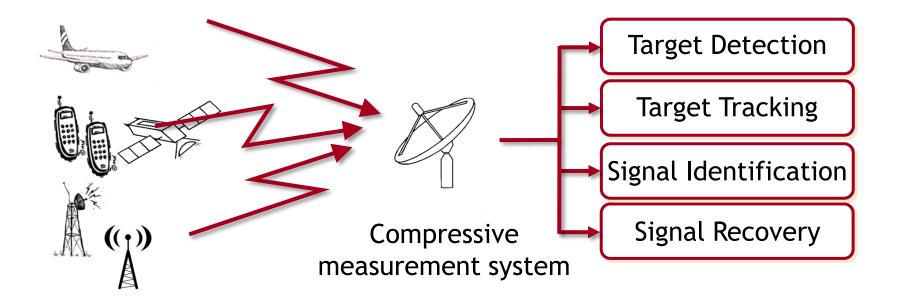
Mark A. Davenport

Stanford University Department of Statistics



Compressive Signal Processing

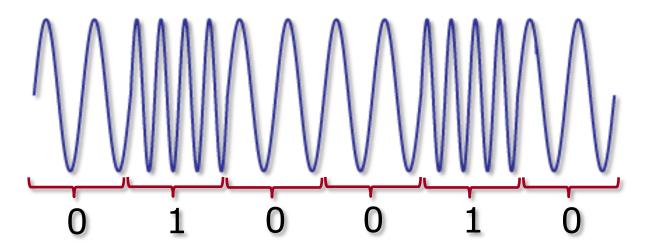
Random measurements are *information scalable*



When and how can we directly solve signal processing problems directly from compressive measurements?

Example: FM Signals

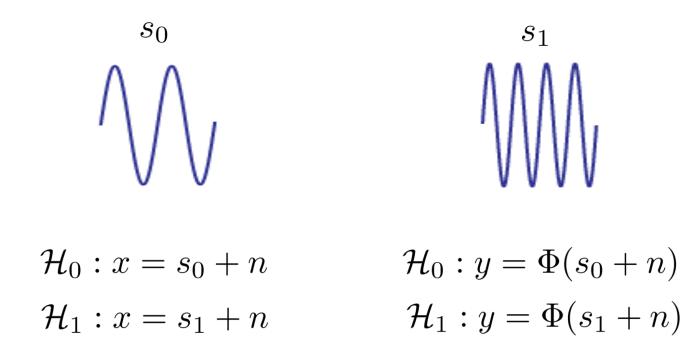
- Can we directly recover a *baseband voice signal* without recovering the modulated waveform?
- Suppose we have compressive measurements of a digital communication signal (FSK modulated)



• Can we directly recover the encoded *bitstream* without first recovering the measured waveform?

Compressive Likelihood-Ratio Test

- Suppose that we are synchronized and know the exact carrier frequencies of the signal
- Window measurements according bit-intervals



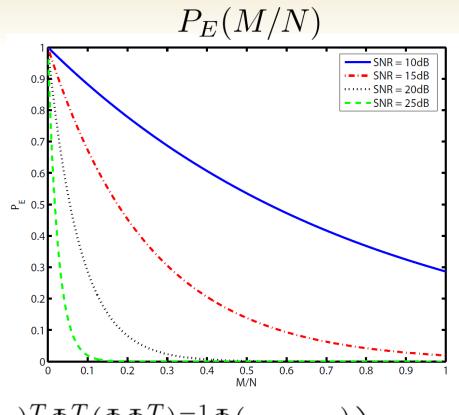
Compressive Classification

 Matched filter (in colored noise)

$$t_0 = s_0^T \Phi^T (\Phi \Phi^T)^{-1} y$$

$$t_1 = s_1^T \Phi^T (\Phi \Phi^T)^{-1} y$$

 Expected performance given by



$$P_D(\alpha) = Q \left(Q^{-1}(\alpha) - \frac{(s_0 - s_1)^T \Phi^T (\Phi \Phi^T)^{-1} \Phi(s_0 - s_1)}{\sigma} \right)$$
$$\approx Q \left(Q^{-1}(\alpha) - \sqrt{\frac{M}{N}} \frac{|s_0 - s_1||_2}{\sigma} \right)$$
ID. Boufouros, Wakin, and Baraniuk - 2010

Compressive Detection

- $\begin{aligned} \mathcal{H}_0 : x &= n & \mathcal{H}_0 : y &= \Phi n \\ \mathcal{H}_1 : x &= s + n & \mathcal{H}_1 : y &= \Phi(s + n) \\ \langle x, s \rangle \geqslant \gamma & s^T \Phi^T (\Phi \Phi^T)^{-1} y \geqslant \widetilde{\gamma} \end{aligned}$
- If Φ is an orthoprojector, then

$$s^T \Phi^T (\Phi \Phi^T)^{-1} y = \sqrt{\frac{N}{M}} s^T \Phi^T y = \sqrt{\frac{N}{M}} \langle y, \Phi s \rangle$$

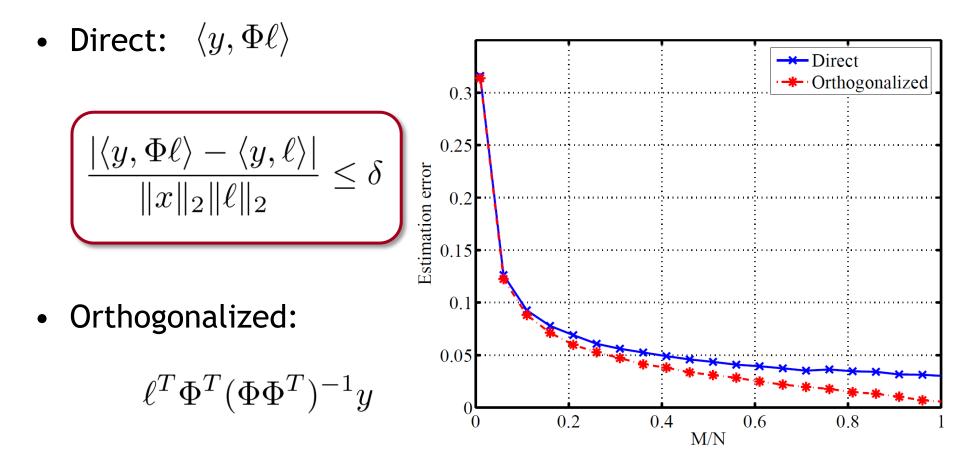
• ROC given by

$$P_D(\alpha) = Q\left(Q^{-1}(\alpha) - \frac{\|\Phi s\|_2}{\sigma}\right) \approx Q\left(Q^{-1}(\alpha) - \sqrt{\frac{M}{N}}\frac{\|s\|_2}{\sigma}\right)$$

[D, Boufounos, Wakin, and Baraniuk - 2010]

Compressive Estimation

• Suppose we wish to estimate $\langle x, \ell \rangle$ from $y = \Phi(x+n)$.

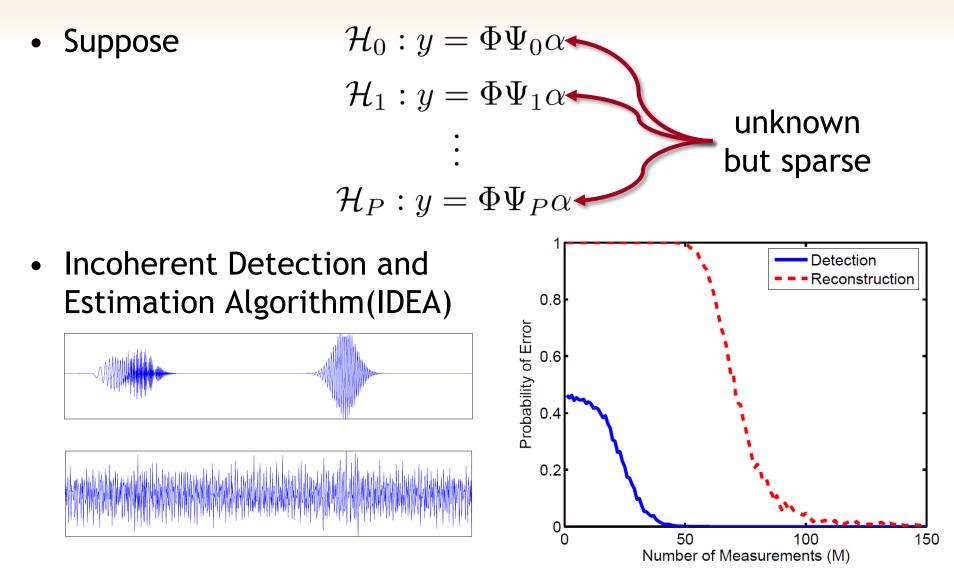


[D, Boufounos, Wakin, and Baraniuk - 2010]

Extreme Sparsity

- In all these examples, we essentially have S = 1.
- All the information we need to solve our problem lies in the 1-dimensional subspace spanned by
 - s in the case of detection
 - $s_0 s_1$ in the case of classification
 - ℓ in the case of estimation
- We have to make a lot of assumptions for these models to be relevant...
- Can these ideas be generalized?

Sparse Signal Classification



[Duarte, D, Wakin, and Baraniuk - 2006]

Matched Filters

 We may know what signals we are looking for, but we may not know where to look

$$H_j: x = s_j(t - \theta_j) + n$$

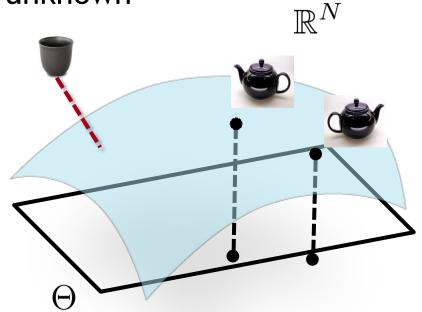
• Elegant solution:

Matched Filter
Compute
$$\langle x, s_j(t - \theta_j) \rangle$$
 for all θ_j

Challenge: Modify the compressive LRT to accommodate unknown parameters

Matched Filter Geometry

- Detection/classification with S unknown articulation parameters $\hfill =$
- Images are points in \mathbb{R}^N
- As template articulation parameters change, points trace out an S-dimensional manifold



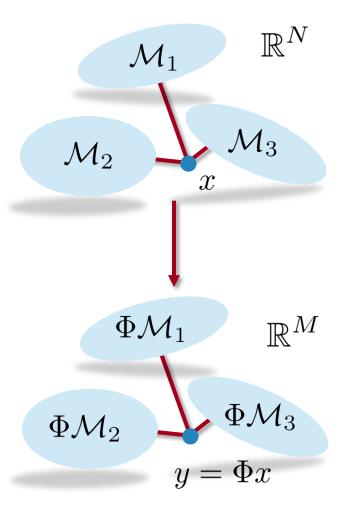
- Classify by finding closest target template to data
- Matched filter based on generalized likelihood ratio test
 closest manifold search

The Smashed Filter

- Compressive manifold classification with GLRT
 - nearest-manifold classifier
 - manifolds classified are now $\Phi \mathcal{M}_j$

- To stably embed ${\boldsymbol{K}}$ manifolds
 - dimension S
 - condition number 1/ au
 - volume V

 $M = O(KS \log(NV/\tau))$



Smashed Filter - Experiments

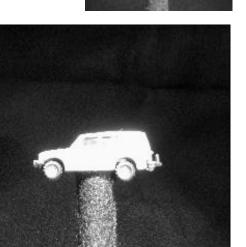
- 3 image classes
 - T-72 tank
 - schoolbus
 - SUV
- Imaged using single-pixel camera with
 - unknown shift
 - unknown rotation





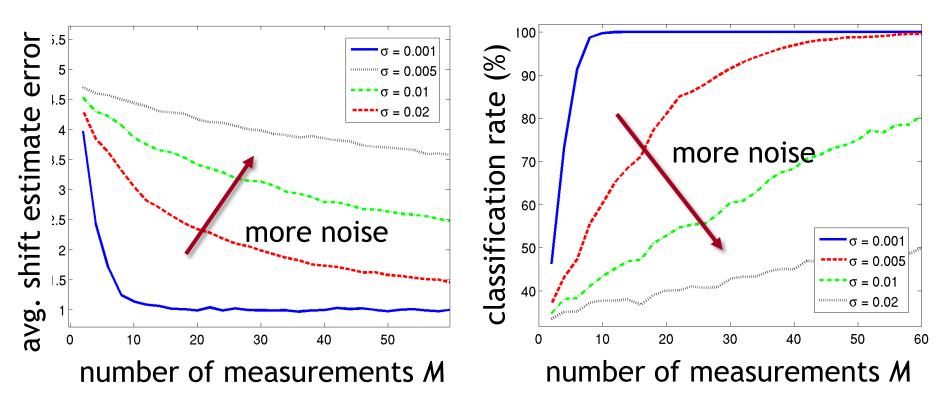






Smashed Filter

- Random shift and rotation (S = 3 dim. manifold)
- AWG noise added to measurements
- Goals: identify most likely shift/rotation parameters identify most likely class



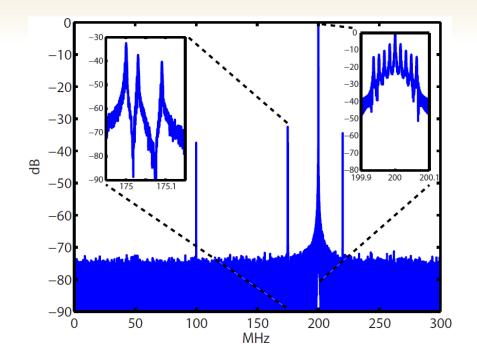
[D, Duarte, Wakin, Laska, Takhar, Kelly, and Baraniuk - 2007]

Putting It All Together

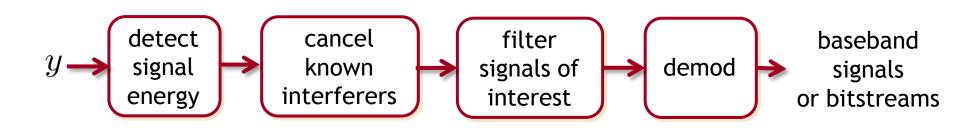
Compressive Radio Receivers

Example Scenario

- 300 MHz bandwidth
- 5 FM signals (12 kHz)
- TV station interference
- Acquire compressive measurements at 30 MHz (20 x undersampled)



We must simultaneously solve several problems

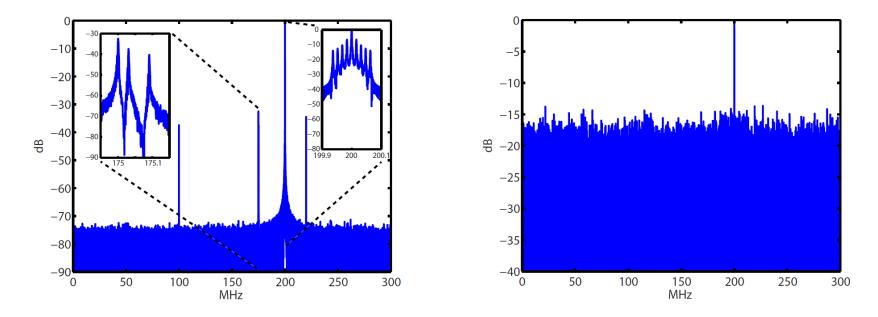


Energy Detection

We need to identify where in frequency the important signals are located

Comprressive Estimation: correlate with projected tones

 $\widehat{F}(k) = |\langle \Phi \cos(2\pi f_k t), y \rangle|$

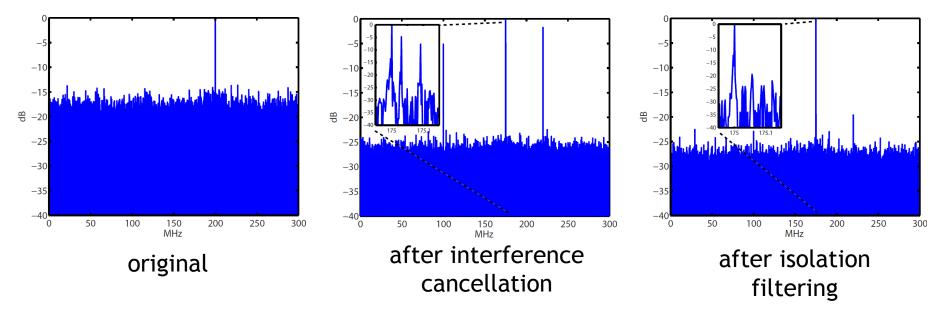


Filtering

If we have multiple signals, must be able to filter to isolate and cancel interference

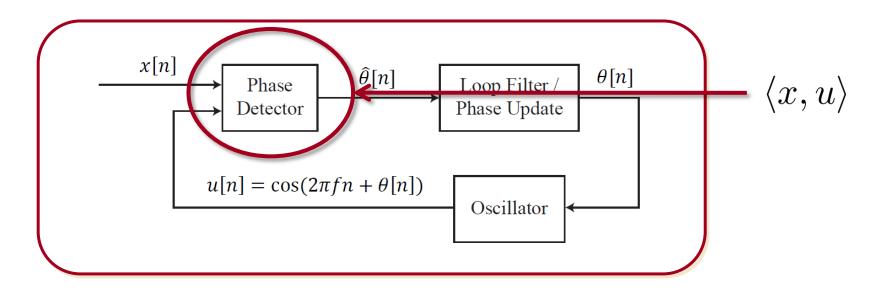
$$P = I - \Phi S (\Phi S)^{\dagger}$$

S: Discrete prolate spheroidal sequences



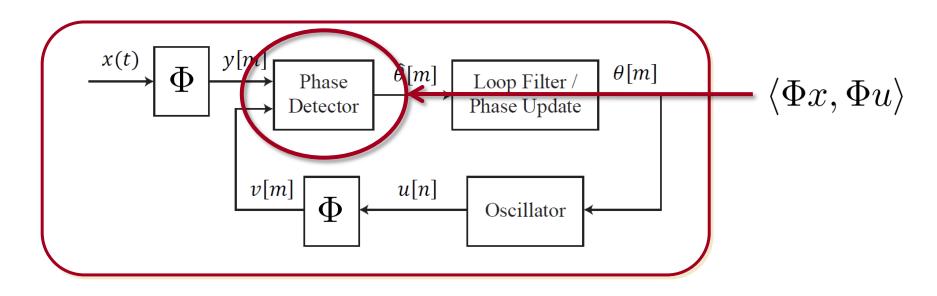
Unsynchronized Demodulation

We can use a phase-locked-loop (PLL) to track deviations in frequency by directly operating on compressive measurements



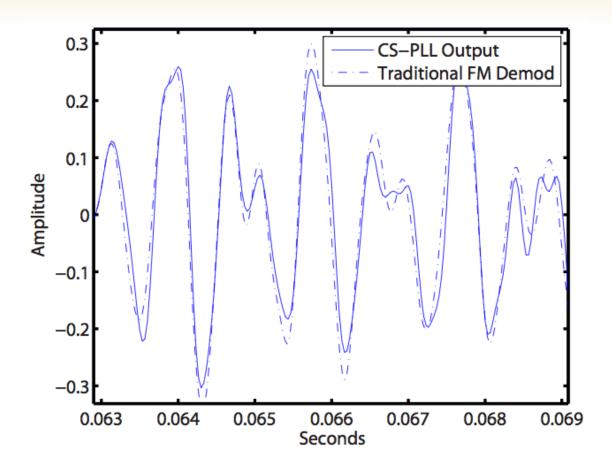
Unsynchronized Demodulation

We can use a phase-locked-loop (PLL) to track deviations in frequency by directly operating on compressive measurements



We can directly demodulate signals from compressive measurements *without recovery*

Compressive Domain Demodulation



CS-PLL with 20x undersampling

Summary

Compressive signal processing

- integrates sensing, compression, processing
- exploits signal sparsity/compressibility
- enables new sensing modalities, architectures, systems
- exploits randomness at many levels
- Why CSP works: preserves information in signals with concise geometric structure sparse signals | manifolds | low-dimensional models
- Information scalability for compressive inference
 - compressive measurements ~ sufficient statistics
 - much less computation required than for recovery